

Algorithms of Scientific Computing

FFT on Real Data

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DFT and Symmetry – Agenda

INPUT TRANSFORM

real symmetry $f_n \in \mathbb{R}$ \rightarrow Real DFT (RDFT)

even symmetry $f_n = f_{-n}$ \rightarrow Discrete Cosine Transform (DCT)

odd symmetry $f_n = -f_{-n}$ \rightarrow Discrete Sine Transform (DST)

“QUARTER-WAVE” INPUT TRANSFORM

even symmetry $f_n = f_{-n-1}$ \rightarrow QW-DCT

odd symmetry $f_n = -f_{-n-1}$ \rightarrow QW-DST

Real-valued DFT (RDFT)

Consider real-valued input data $f_n \in \mathbb{R}$, i.e.: $f_n^* := \bar{f}_n = f_n$, then:

$$F_k = \frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n e^{-i2\pi nk/N} = \frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n \left(\cos\left(\frac{2\pi nk}{N}\right) - i \sin\left(\frac{2\pi nk}{N}\right) \right).$$

Properties:

- $\operatorname{Re}\{F_k\} = \frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n \cos\left(\frac{2\pi nk}{N}\right)$, $\operatorname{Im}\{F_k\} = -\frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n \sin\left(\frac{2\pi nk}{N}\right)$
- only N independent, real coefficients necessary, since:

$$F_k^* = \frac{1}{N} \sum f_n^* \left\{ \omega_N^{-nk} \right\}^* = \frac{1}{N} \sum f_n \omega_N^{-n(-k)} = F_{-k}$$

$$\begin{aligned} \text{Recall: } \left\{ \omega_N^{-nk} \right\}^* &= \left\{ e^{-i2\pi nk/N} \right\}^* = \cos(-2\pi nk/N) - i \sin(-2\pi nk/N) \\ &= \cos(2\pi nk/N) + i \sin(2\pi nk/N) = \dots = \omega_N^{nk} \end{aligned}$$

Real DFT (2)

Map N values to N coefficients (and vice versa):

$$\left\{ f_{-\frac{N}{2}+1}, \dots, f_0, \dots, f_{\frac{N}{2}} \right\}$$

DFT \Downarrow \Uparrow IDFT

$$\left\{ F_0, \operatorname{Re}\{F_1\}, \operatorname{Im}\{F_1\}, \dots, \operatorname{Re}\{F_{\frac{N}{2}-1}\}, \operatorname{Im}\{F_{\frac{N}{2}-1}\}, F_{\frac{N}{2}} \right\}$$

Note: real and imaginary parts of F_{-k} correspond to those of F_k

Real DFT (3)

Situation:

- only N real input values (as all N imaginary parts are 0)
- **only N independent, real output values** (coefficient components)
due to symmetry $F_{-k} = F_k^*$

Wanted → new transformation:

N real input values → N distinct real coefficient components

Hence: Insert symmetry $F_{-k} = F_k^*$ in IDFT!

Real DFT (4)

Definition of “**Real discrete Fourier transform**” (RDFT)

- formulation 1

$$F_k = \frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n \left(\cos \left(\frac{2\pi nk}{N} \right) - i \sin \left(\frac{2\pi nk}{N} \right) \right), k = 0, \dots, \frac{N}{2}$$

- formulation 2

$$\operatorname{Re}\{F_k\} = \frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n \cos \left(\frac{2\pi nk}{N} \right), k = 0, \dots, \frac{N}{2}$$

$$\operatorname{Im}\{F_k\} = -\frac{1}{N} \sum_{n=-\frac{N}{2}+1}^{\frac{N}{2}} f_n \sin \left(\frac{2\pi nk}{N} \right), k = 1, \dots, \frac{N}{2} - 1$$

Inverse Real DFT

Goal: compute a real representation of the DFT (i.e., no $e^{-i2\pi nk/N}$).

We start with an input vector (Fourier coefficients) of **length $2N$** , then:

$$\begin{aligned}
 f_n &= \sum_{k=-N+1}^N F_k e^{i2\pi nk/2N} \\
 &= F_0 + \sum_{k=1}^{N-1} \left(F_k e^{i2\pi nk/2N} + F_{-k} e^{-i2\pi nk/2N} \right) + F_N e^{i2\pi nN/2N} \\
 &= F_0 + \sum_{k=1}^{N-1} \left(F_k e^{i2\pi nk/2N} + \left\{ F_k e^{i2\pi nk/2N} \right\}^* \right) + F_N e^{i\pi n} \\
 &= F_0 + 2 \sum_{k=1}^{N-1} \operatorname{Re} \left\{ F_k e^{i2\pi nk/2N} \right\} + F_N e^{i\pi n} \\
 &= F_0 + 2 \sum_{k=1}^{N-1} \left(\operatorname{Re}\{F_k\} \cos\left(\frac{\pi nk}{N}\right) - \operatorname{Im}\{F_k\} \sin\left(\frac{\pi nk}{N}\right) \right) + F_N \cos(\pi n)
 \end{aligned}$$

Inverse Real DFT (2)

Set $a_k := 2 \operatorname{Re}\{F_k\}$ and $b_k := -2 \operatorname{Im}\{F_k\}$ (but $a_0 := \operatorname{Re}\{F_0\}$ and $a_N := \operatorname{Re}\{F_N\}$) to get :

$$f_n = a_0 + \sum_{k=1}^{N-1} \left(a_k \cos\left(\frac{\pi nk}{N}\right) + b_k \sin\left(\frac{\pi nk}{N}\right) \right) + a_N \cos(\pi n)$$

“Real inverse discrete Fourier transform”

Using the (real-valued) formula for F_k :

$$a_k = \frac{1}{N} \sum_{n=-N+1}^N f_n \cos\left(\frac{\pi nk}{N}\right), \quad b_k = \frac{1}{N} \sum_{n=-N+1}^N f_n \sin\left(\frac{\pi nk}{N}\right)$$

but: $a_0 = \frac{1}{2N} \sum \dots$ and $a_N = \frac{1}{2N} \sum \dots$

Inverse Real DFT – Compare with Textbooks

Alternate (probably more frequent) notation in textbooks:

Set $a_k := 2 \operatorname{Re}\{F_k\}$ and $b_k := -2 \operatorname{Im}\{F_k\}$ to get

$$f_n = \frac{1}{2}a_0 + \sum_{k=1}^{N-1} \left(a_k \cos\left(\frac{\pi nk}{N}\right) + b_k \sin\left(\frac{\pi nk}{N}\right) \right) + \frac{1}{2}a_N \cos(\pi n)$$

Using the (real-valued) formula for F_k :

$$a_k = \frac{1}{N} \sum_{n=-N+1}^N f_n \cos\left(\frac{\pi nk}{N}\right), \quad b_k = \frac{1}{N} \sum_{n=-N+1}^N f_n \sin\left(\frac{\pi nk}{N}\right)$$

Differences:

- no extra definition of a_0 and a_N
- but factors $\frac{1}{2}$ in formula for $f_n \rightarrow$ not as nicely related to interpolation

Real-valued Trigonometric Interpolation

Interpretation of the real DFT as an interpolation problem:

- $2N$ ansatz functions:

$$\begin{aligned} g_k(x) &:= \cos(kx) & k = 0, \dots, N \\ h_k(x) &:= \sin(kx) & k = 1, \dots, N-1 \end{aligned}$$

- $2N$ supporting points: $x_n := \frac{2\pi n}{2N} = \frac{\pi n}{N} \quad n = -N+1, \dots, N$
- $2N$ interpolation conditions:

$$f_n = a_0 + \sum_{k=1}^{N-1} \left(a_k \cos\left(\frac{\pi nk}{N}\right) + b_k \sin\left(\frac{\pi nk}{N}\right) \right) + a_N \cos(\pi n)$$

(cmp. exercises)

Fast Real DFT

Computation of a real-valued DFT using complex FFT is inefficient:

- N redundant components computed (symmetry)
- complex arithmetics with lots of real/imaginary parts being 0

Possibilities to improve the efficiency:

1. compute two real DFTs from one complex FFT
2. compute a real DFT of length $2N$ from one complex FFT of length N
3. “compact” real FFT – use symmetry of the data directly in the algorithm

Two Real DFTs from one complex FFT

Idea: for real-valued g_n and h_n , compute DFT of $f_n := g_n + ih_n$:

$$F_k = \frac{1}{N} \sum_n (g_n + ih_n) \omega_N^{-nk}$$

Comparison with coefficients G_k and H_k of the two real DFTs:

$$G_k = \frac{1}{N} \sum_n g_n \omega_N^{-nk} \quad H_k = \frac{1}{N} \sum_n h_n \omega_N^{-nk}$$

Due to linearity of the Fourier transform:

$$F_k = G_k + iH_k$$

Two Real DFTs from one complex FFT (2)

Since g_n and h_n are real data, we have the following symmetry:

$$G_k = G_{-k}^* \quad H_k = H_{-k}^* .$$

Hence, we get for F_{-k}^* :

$$F_{-k}^* = (G_{-k} + iH_{-k})^* = (G_{-k}^* + i^* H_{-k}^*) = G_k - iH_k .$$

Together with $F_k = G_k + iH_k$, we obtain

$$G_k = \frac{1}{2} (F_k + F_{-k}^*) \quad \text{and} \quad H_k = -\frac{i}{2} (F_k - F_{-k}^*) .$$

Two real DFTs from one complex FFT – Algorithm

Algorithm to compute two real DFTs:

- (1) set $f_n := g_n + ih_n$
- (2) compute F_k from FFT (using a library, e.g.)
- (3) compute G_k and H_k according to

$$G_k = \frac{1}{2} (F_k + F_{-k}^*) \quad \text{and} \quad H_k = -\frac{i}{2} (F_k - F_{-k}^*)$$

⇒ “half” the costs compared to using complex FFT

but: additional operations for pre- and postprocessing

Real DFT of length $2N$ from complex FFT of length N

Compute DFT of a real-valued vector (f_{-N+1}, \dots, f_N) :

$$F_k = \frac{1}{2N} \sum_{-N+1}^N f_n \omega_{2N}^{-nk} \quad \text{for } k = -\frac{N}{2} + 1, \dots, \frac{N}{2}$$

Split up in $g_n := f_{2n}$ and $h_n := f_{2n-1}$; leads to butterfly scheme:

$$\begin{aligned} F_k &= \frac{1}{2} \left(G_k + \omega_{2N}^k H_k \right), \\ F_{k \pm N} &= \frac{1}{2} \left(G_k - \omega_{2N}^k H_k \right) \end{aligned}$$

for $k = -\frac{N}{2} + 1, \dots, \frac{N}{2}$, respectively.

Real $2N$ -DFT from complex N -FFT

Now: compute G_k and H_k (two real DFTs) from one complex FFT
 \rightarrow applied to $z_n := g_n + ih_n = f_{2n} + if_{2n-1}$:

$$G_k = \frac{1}{2} (Z_k + Z_{-k}^*) \quad \text{and} \quad H_k = -\frac{i}{2} (Z_k - Z_{-k}^*)$$

Combine both schemes to:

$$\begin{aligned} F_k &= \frac{1}{4} Z_k (1 - i\omega_{2N}^k) + \frac{1}{4} Z_{-k}^* (1 + i\omega_{2N}^k), & k = 0, \dots, \frac{N}{2} \\ F_{k+N} &= \frac{1}{4} Z_k (1 + i\omega_{2N}^k) + \frac{1}{4} Z_{-k}^* (1 - i\omega_{2N}^k), & k = -\frac{N}{2} + 1, \dots, 0 \end{aligned}$$

Real $2N$ -DFT from complex N -FFT – Algorithm

Algorithm for a real $2N$ -DFT:

- (1) set $z_n := f_{2n} + if_{2n-1}$
- (2) compute Z_k from FFT applied on z_n (using a library, e.g.)
- (3) compute F_k according to

$$F_k = \frac{1}{4}Z_k \left(1 - i\omega_{2N}^k\right) + \frac{1}{4}Z_{-k}^* \left(1 + i\omega_{2N}^k\right), \quad k = 0, \dots, \frac{N}{2}$$

$$F_{k+N} = \frac{1}{4}Z_k \left(1 + i\omega_{2N}^k\right) + \frac{1}{4}Z_{-k}^* \left(1 - i\omega_{2N}^k\right), \quad k = -\frac{N}{2} + 1, \dots, 0$$

⇒ Complexity determined by complex N -FFT

plus: additional operations for pre- and postprocessing

Compact Real FFT

Compute DFT of a real-valued vector (f_{-N+1}, \dots, f_N) :

$$F_k = \frac{1}{2N} \sum_{-N+1}^N f_n \omega_{2N}^{-nk} \quad \text{for } k = 0, \dots, N$$

Split up in $g_n := f_{2n}$ and $h_n := f_{2n-1}$; leads to butterfly scheme:

$$\begin{aligned} F_k &= \frac{1}{2} \left(G_k + \omega_{2N}^k H_k \right) & \text{for } k = 0, \dots, \frac{N}{2}, \\ F_{k+N} &= \frac{1}{2} \left(G_k - \omega_{2N}^k H_k \right) & \text{for } k = -\frac{N}{2} + 1, \dots, 0. \end{aligned}$$

G_k and H_k are coefficients of a real-valued DFT of length N ; hence:

$$G_k = G_{-k}^* \quad \text{and} \quad H_k = H_{-k}^* \quad \text{for } k = 0, \dots, \frac{N}{2} - 1$$

Compact Real-valued FFT (2)

Use symmetry of G_k and H_k for the computation of F_k :

$$\begin{aligned}
 F_k &= \frac{1}{2} \left(G_k + \omega_{2N}^k H_k \right) && \text{für } k = 0, \dots, \frac{N}{2}, \\
 F_{N-k} &= \frac{1}{2} \left(G_{-k} - \omega_{2N}^{-k} H_{-k} \right) \\
 &= \frac{1}{2} \left(G_k - \omega_{2N}^k H_k \right)^* && \text{for } k = 0, \dots, \frac{N}{2} - 1
 \end{aligned}$$

⇒ Computation of F_k (for $k = 0, \dots, N$) reduced to the computation of G_k and H_k (for $k = 0, \dots, \frac{N}{2}$, respectively).

“Edson’s algorithm” (1968)